

Midterm #1 Solutions – EECS 145L Fall 1999

- 1a** The op-amp equation is $V_0 = A(V_+ - V_-)$
 If V_0 is finite and A is infinite, then $V_+ = V_-$ (virtual short rule)
 $V_- = V_+ = V_2 R_2 / (R_1 + R_2)$
 Since no current flows in or out of the op-amp inputs

$$\frac{V_+ - V_1}{R_1} + \frac{V_+ - V_0}{R_2} = \frac{V_+(R_1 + R_2) - V_1 R_2 - V_0 R_1}{R_1 R_2} = \frac{V_2 R_2 - V_1 R_2 - V_0 R_1}{R_1 R_2} = 0$$

$$\frac{V_2 - V_1}{R_1} = \frac{V_0}{R_2} \quad G_{\pm} = V_0 / (V_2 - V_1) = R_2 / R_1$$
 [5 points off for correct setup followed by algebraic errors]
 [10 points off for trying without using the virtual short rule]
 [15 points off for missing or incorrect setup]

- 1b** To determine common mode gain, set $V_1 = V_2$ in the equation $\frac{V_2 - V_1}{R_1} = \frac{V_0}{R_2}$, and we have
 $V_0 = 0$ and $G_c = 0$

2 $4 \text{ nV Hz}^{-1/2} \sqrt{f} = 1.29 \times 10^{-10} \text{ }^{-1/2} \text{ Hz}^{-1/2} \sqrt{R} f$
 $\sqrt{R} = 40 / 1.29 \quad R = 961$

- 3a** Differential gain 1000, bandwidth 10 kHz
 [3 points off for Gain = 10,000, bandwidth 1 kHz]
- 3b** Output $V_{\text{rms}} = (4 \text{ nV Hz}^{-1/2}) (100 \text{ Hz}^{1/2}) (1000) = 0.4 \text{ mV}$ in 10 kHz
 [2 points off for 0.4 μV]
- 3c** We want a Butterworth low-pass filter with a gain of 0.99 at 1 kHz and 0.01 at 2 kHz. Look at the 0.99 and 0.01 columns of the equation sheet to find a filter order where the f/f_c ratios are 2 or a bit less. Order 6 gives $2.154/0.723 = 2.98$ (too high). Order 8 gives $1.778/0.784 = 2.27$ (still too high). Order 10 gives $1.585/0.823 = 1.93$ (OK).
 The order is 10 and the corner frequency is $1/0.823 = 1.22 \text{ kHz}$. (order 12 also accepted)
 [5 points off for not giving corner frequency]
 [2 points off for not specifying low pass]
 [2 points off for order 8]
 After amplification and filtering, the output noise would be $V_{\text{rms}} = (4 \text{ nV Hz}^{-1/2}) \sqrt{1.22 \text{ kHz}} (1000) = (4 \text{ nV Hz}^{-1/2}) (34.9 \text{ Hz}^{1/2}) (1000) = 0.139 \text{ mV}$ in 10 kHz
 [4 points off if output noise not given]
 So the filtering reduced the output noise from $\pm 0.4 \text{ mV}$ to $\pm 0.14 \text{ mV}$
- 3d** The best way to reduce the 60 Hz interference from the middle of a band of frequencies of interest is to use a notch filter. The common mode rejection ratio of 60 dB means that the common mode gain is $1000/1000 = 1$. So the instrumentation amplifier 60 Hz output will be $\pm 10 \text{ mV}$ from a common mode input of $\pm 10 \text{ mV}$. The output due to a differential 60 Hz interference is $(\pm 0.01 \text{ mV}) (1000) = \pm 10 \text{ mV}$. In the worst case, these are in phase, producing a total of $\pm 20 \text{ mV}$. A notch filter can reduce this total by a factor of typically 30, to $\pm 0.7 \text{ mV}$.

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[Any value between 0.1 and 2 mV was accepted for full credit]

[3 points off for including one 10 mV and not the other]

[6 points off if input noise not given]

[10 points off for using a HPF, which removes the important earthquake frequencies below 60 Hz]

- 3e** Seismometer followed by instr amp of gain 1000, followed by notch filter (60 Hz) followed by low pass filter of order 10 and corner frequency 1.22 kHz. [OK to reverse order of low pass and notch filters]

145L midterm #1 grade distribution:

		maximum score =	100	
		average score =	89.4	
Problem		66-70	0	
1	32.5 (35 max)	71-75	1	
2	9.7 (10 max)	76-80	4	C
3	47.3 (55 max)	81-85	1	
		86-90	3	B
		91-95	4	B
		96-100	6	A